#### **Features**

- Digital angular rate sensor with SPI interface
- Angular rate measurement around Z-axis (yaw)
- ±300°/sec input range
- Ultra low noise
- Excellent bias instability
- 24 bit angular rate output
- Embedded temperature sensor for on-chip or external temperature compensation
- Built-in Self-Test
- 5V single supply voltage
- Low operating current consumption: 25mA
- CLCC 30 package: 19.6 mm x 11.5 mm x 2.9 mm
- Weight : 2 grams
- REACH and RoHS compatible

## **Applications**

- Precision instrumentation
- Platform stabilization and control
- Unmanned aerial vehicles



#### **General Description**

GYPRO<sup>®</sup> product line is an established family of Micro-Electro-Mechanical Systems (MEMS) angular rate sensor specifically designed for demanding applications.

The MEMS transducer is manufactured using Tronics proprietary vacuum wafer-level packaging technology based on micro-machined thick single crystal silicon.

The integrated circuit (IC) provides a stable primary antiphase vibration of the 'drive' proof masses, thanks to electrostatic comb drives. When the sensor is subjected to a rotation, the Coriolis force acts on the 'sense' proof masses and forces them into a secondary anti-phase movement perpendicular to the direction of drive vibration, which is itself counter-balanced by electrostatic forces. The sense closed loop operates as an electromechanical  $\Sigma\Delta$  modulator providing a digital output. This output is finally demodulated using the drive reference signal.

The sensor is factory calibrated and compensated for temperature effects to provide high-accuracy digital output over a broad temperature range.

Raw data output can be also chosen to enable customer-made compensations.

Part Number	G2300	G2310	G3300	G4300
Description	Standard configuration	Low delay configuration	Improved vibration tolerance & ultra-low delay configuration	Improved bias thermal stability & reduced dimensions
Latency	40ms	2ms	1ms	1ms
Vibration range	4g rms	4g rms	8g rms	8g rms
Bandwidth	100Hz	>200Hz	>200Hz	>200Hz
Datarate	200Hz	1700Hz	1800Hz	1800Hz
Angular Random Walk	0.14°/√hr	0.14°/Vhr	0.15°/Vhr	0.10°/√hr
Size (L x l x h)	19 x 11 x 3mm	19 x 11 x 3mm	19 x 11 x 4mm	12 x 12 x 6mm
Package	CLCC 30	CLCC 30	CLCC 30	28 pins J-Lead

# **GYPRO<sup>®</sup> Product references**

#### Disclaimer

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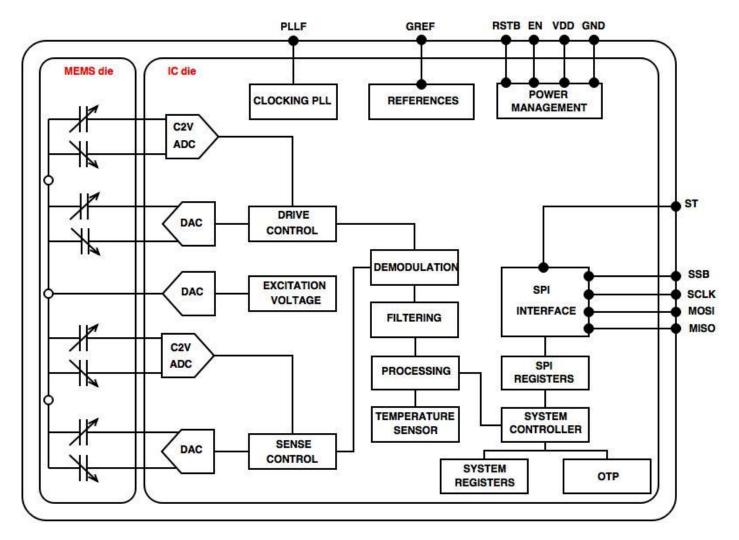
## Contents

Featu	ures	1
Appli	ications	1
Gene	eral Description	1
GYPF	RO® Product references	1
Discl	aimer	1
Blo	ck diagram	3
Ove	erall Dimensions	3
1.	Specifications	4
2.	Maximum Ratings	6
3.	Typical performances	7
4.	Interface	9
4.1.	. Pinout, sensitive axis identification	9
4.2.	. Application circuit	9
4.3.	. Input/Output Pin Definitions	
5.	Soldering Recommendations	
6.	Digital SPI interface	
6.1.	. Electrical and Timing Characteristics	
6.2.	. SPI frames description	14
6.3.	. Angular rate readings	
6.4.	. Temperature readings	
6.5.	. Advanced use of SPI registers	
7.	Angular rate calibration procedure	
7.1.	. Algorithm overview	
7.2.	. Programming of the new coefficients	
7.3.	. Switch to uncompensated data output	
8.	Temperature Sensor Calibration Procedure	19
8.1.	. Temperature sensor calibration model	19
8.2.	. Recommended Procedure	
9.	Device Identification/Ordering information	20
9.1.	. Device identification	20
9.2.	. Ordering information	20
10.	Internal construction and Theory of Operation	
11.	Available Tools and Resources	

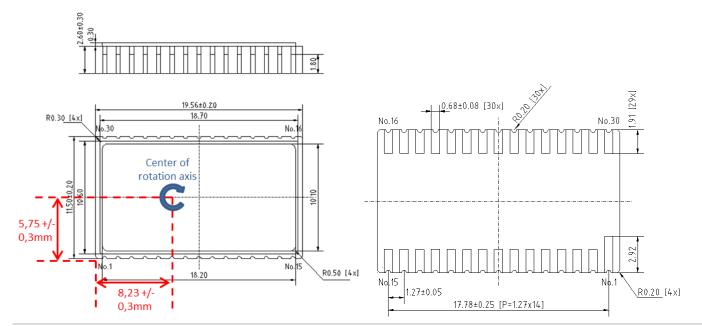
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# tronics GYPRO2300 Datasheet

#### **Block diagram**



#### **Overall Dimensions**



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# 1. Specifications

Unless specified in brackets, GYPRO2300LD characteristics are the same as GYPRO2300.

Parameter	Unit	Тур.	Max	Notes
Measurement Ranges				
Input range*	°/s	±300	±350	Electronic clamping is applied and sensors will saturate before ±500°/s.
Temperature range *	°C	-40 t	0 +85	
Bias				
Bias instability	°/h	0.8		Lowest point of Allan variance curve at room temperature.
Bias in-run (short term) stability	°/h	30		Standard deviation of the 1 second filtered output over 1 hour at room temperature, after 30 min of stabilization.
Residual Bias Temperature Error, calibrated *	°/s	0.05	0.2	Peak to peak deviation of the bias over the specified temperature range after warm-up. Factory calibration is performed in test socket following one stabilization thermal cycle and while sensor is powered up. As printed circuit board reflow soldering may cause shifts in bias temperature variations, it may be necessary to do an on-board calibration after soldering, depending on applications requirements.
Bias run to run repeatability	°/h	10		Standard deviation of 7 bias measurements at 30°C that occurs between seven runs of operation with 30 minutes power off between each run.
Vibration rectification coefficient	°/h/g²	30		Bias rectification under vibration, overall level 4g rms, according to Figure 11.
Scale Factor (tested on ±300°	/s range)			
Scale Factor *	LSB/°/s	10 000		Nominal scale factor.
Residual Scale factor Temperature Error, calibrated *	%	0.2	0.75	Peak to peak deviation of the scale factor over the specified temperature range.
Scale Factor run to run repeatability	ppm	100		Standard deviation of 7 scale factor measurements at 30°C that occurs between seven runs of operation with 30 minutes power off between each run.
Scale factor non linearity*	ppm	70	500	Maximum deviation of the output from the expected value using a best fit straight line, at room temperature.
Noise				
RMS Noise [1-100Hz] *	°/s	0.02	0.05	RMS noise level in the band [1-100Hz], obtained by integrating the power spectral density of the sensor output between 1 and 100Hz at zero rate and room temperature.
Angular random walk	°/√h	0.14		-1/2 slope of Allan variance curve at room temperature.
Frequency response				
Bandwidth	Hz	100 (>200)		Defined as the frequency for which attenuation is equal to -3dB.
Resonant Frequency	Hz	3300 t	o 3950	Drive resonant frequency of the sensor, at room temperature.



Hz ms s /h/g ms nrad	190 to (1560 to (2) 0.8 18 10		Refresh rate of the output data at room temperature.   Group delay of the filtering chain.   Time interval between application of power on and the availability of an output signal (at least 90% of the input rate, at room temperature.   Mean value on all axis of output variations under 1g.   Time interval between an impact (half sine 50 g, 6 ms) and the presence of a usable output of the sensor.
s /h/g ms	(2) 0.8 18		Time interval between application of power on and the availability of an output signal (at least 90% of the input rate, at room temperature. Mean value on all axis of output variations under 1g. Time interval between an impact (half sine 50 g, 6 ms) and the
/h/g ms	18		availability of an output signal (at least 90% of the input rate, at room temperature. Mean value on all axis of output variations under 1g. Time interval between an impact (half sine 50 g, 6 ms) and the
ms			Time interval between an impact (half sine 50 g, 6 ms) and the
ms			Time interval between an impact (half sine 50 g, 6 ms) and the
-	10		
nrad			·
nrad			
		16	Misalignment between the sensitive axis and the normal to the package bottom plane, by design.
°C	-55 to	+100	
%	<98		
	1		Unlimited floor life out of the bag (hermetic package).
ms	50	6	Half sine.
ms	2000	0.3	
g <sub>rms</sub>	L	ļ	See Figure 11
g <sub>rms</sub>	2	0	
V	4.75 to	o 5.25	
mA	25		
μΑ	1	<5	Power down mode is activated by switching EN pin to GND.
/h/V	40		
SB/°C	20		Temperature sensor is not factory-calibrated.
LSB	2000		Temperature sensor is not factory-calibrated.
Hz	6		
	°C %    ms   ms grms grms V mA µA /h/V SB/°C LSB	°C -55 to % <98 1   ms 50   ms 2000 grms 2 grms 2 V 4.75 to mA 25 μA 1 /h/V 40 5B/°C 20 LSB 2000	°C -55 to +100   % <98

**Table 1 Specifications** 

\* 100% tested in production.

\*\* Unless otherwise specified, max values are ±3 sigma variation limits from validation test population.

## 2. Maximum Ratings

Stresses at or exceeding the maximum ratings listed below may cause permanent damage to the device, or affect its reliability. Exposure to maximum ratings conditions for extended periods may also affect device reliability.

Functional operation is not guaranteed once stresses exceeding the maximum ratings have been applied.

Parameter	Unit	Min	Max
Supply Voltage	V	-0.5	+7
Electrostatic Discharge (ESD) protection, any pin, Human Body Model	kV		±2
Storage temperature range	°C	-55	+100
Shock survival	g		2000
Vibrations survival, 20-2000Hz	grms		20
Ultrasonic cleaning		Not allowed	

**Table 2 Maximum ratings** 

# **Caution!**



The product may be damaged by ESD, which can cause performance degradation or device failure! We recommend handling the device only on a static safe work station. Precaution for the storage should also be taken.



The sensor MUST be powered-on *before* any SPI operation, as shown in Figure 1 below. Having the SPI pads, or EN at a high level while VDD is at a low level could damage the sensor, due to ESD protection diodes and buffers.

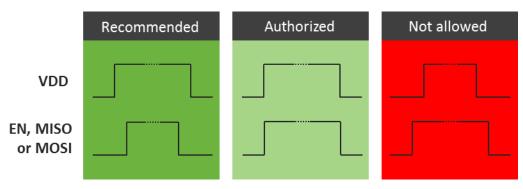


Figure 1 Recommended voltage sequence

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# 3. Typical performances

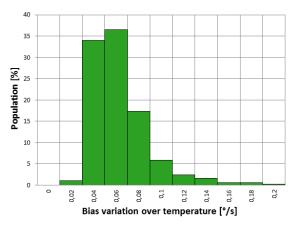


Figure 2 Distribution of bias over temperature

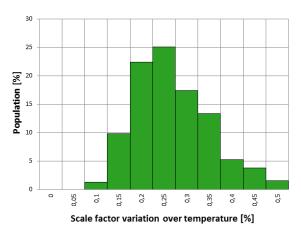


Figure 3 Distribution of scale factor over temperature

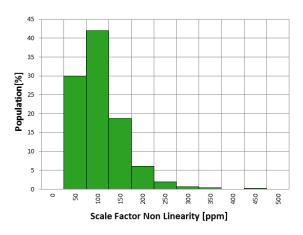


Figure 4 Distribution of scale factor non linearity (RT)

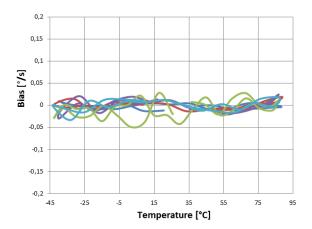


Figure 5 Bias variation over temperature (5 samples)

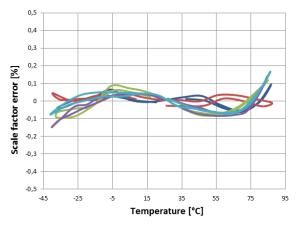


Figure 6 Scale factor variation over temperature (5 samples)

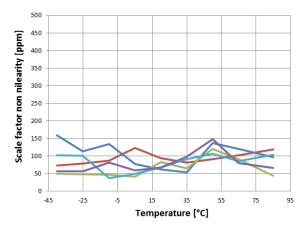
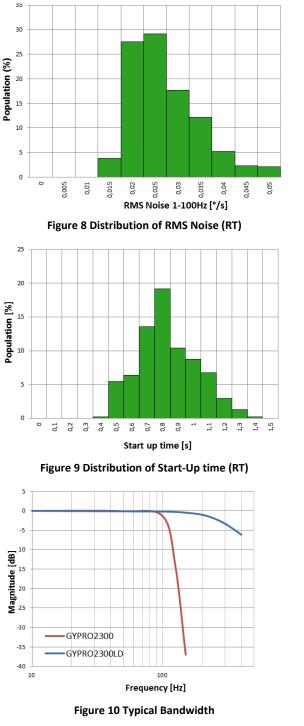
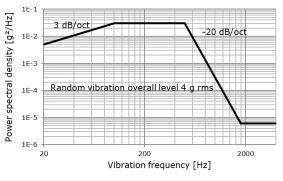


Figure 7 Scale factor non linearity over temperature (5 samples)

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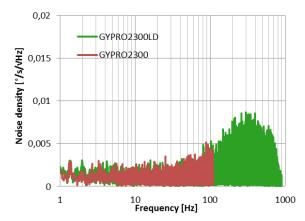


Figure 12 Typical Noise density (RT)

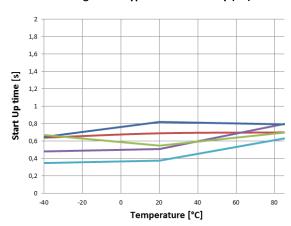


Figure 13 Start-Up Time variation over temperature

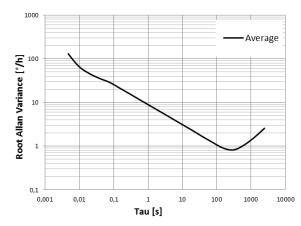


Figure 14 Allan variance (RT)

Figure 11 Operating vibration profile

## 4. Interface

#### 4.1. Pinout, sensitive axis identification

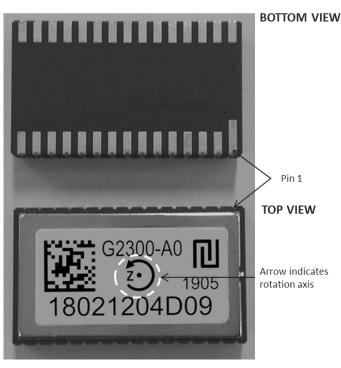
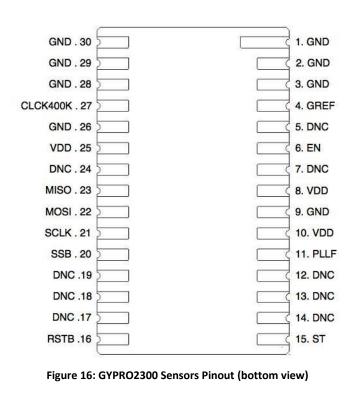
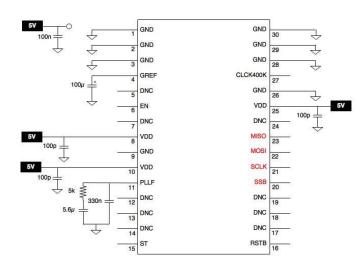


Figure 15: How to locate Pin 1







#### Figure 17: Recommended Application Schematic (top view)

Notes:

- All capacitances of Figure 17 should be placed as close as possible to their corresponding pins, except the 100nF capacitance between VDD and GND, which should be as close as possible to the board's supply input.
- The 100 $\mu$ F filtering capacitance between GREF and GND should have low Equivalent Series Resistance (ESR < 1 $\Omega$ ) and low leakage current (< 6 $\mu$ A). A tantalum capacitor is recommended.
- 5.6μF and 330nF filtering capacitance between PLLF and GND should have a low leakage current (<1μA).</li>

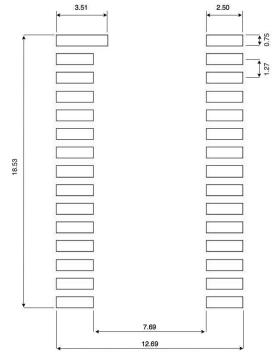


Figure 18: Recommended Pad Layout in mm (top view)

## 4.3. Input/Output Pin Definitions

Pin name	Pin number	Pin type	Pin direction	Pin levels	Function
GND	1, 2, 3, 9, 26, 28, 29, 30	Supply	n/a	0V	Power Ground
VDD	8, 10, 25	Supply	n/a	+5V	Power Supply
GREF	4	Analog	n/a	4.4V	External decoupling pad. MUST be connected to the board's VSS through a 100µF external capacitor, in order to ensure low noise.
EN	6	Digital	Input	VDD with pull- up of 100kΩ	Enable command. Active high.
PLLF	11	Analog	Output	0.8V	External filtering pad. MUST be connected to a filtering stage, described in Figure 17.
ST	15	Digital	Output	VDD	Self-test status. Logic "1" when the sensor is OK.
RSTB	16	Digital	Input	VDD with pull- up of 100kΩ	Reset. Reloads the internal calibration data. Active low
SSB	20	Digital	Input	VDD	Slave Selection signal. Active low
SCLK	21	Digital	Input	VDD	SPI clock signal
MOSI	22	Digital	Input	VDD	Master Output Slave Input signal
MISO	23	Digital	Output	VDD	Master Input Slave Output signal
CLCK400K	27	Digital	Output	VDD	Internal clock
DNC	5, 7, 12, 13, 14, 17, 18, 19, 24				Do Not electrically Connect. These pins provide additional mechanical fixing to the board and should be soldered to an unconnected pad.

Table 3: Pin Functions

Note: The digital pads maximum ratings are GND-0.3V and VDD+0.3V.

## 5. Recommendations

#### 5.1. Soldering

Please note that the reflow profile to be used does not depend only on the sensor. The whole populated board characteristics shall be taken into account.

MEMS components are sensitive to mechanical stress coming from the Printed Circuit Board (PCB) during the soldering reflow. This stress is caused by the mismatch between the Coefficient of Thermal Expansion (CTE) of the ceramic package and the PCB and can affect the Bias temperature variations. In order to achieve the best performance, it is recommended to do an on-board calibration after the soldering of the sensor.

For a better reliability of the soldering, Tronics recommends using Copper-Invar-Copper or ceramic boards. These types of boards have a coefficient of thermal expansion (CTE) close to the CTE of GYPRO2300 package (6.8 ppm/°C).

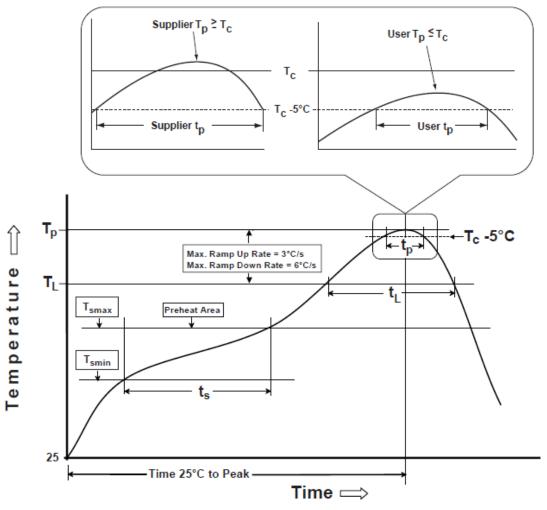


Figure 19: Reflow Profile, according to IPC/JEDEC J-STD-020D.1

Profile Feature	Eutectic Assembly
Time maintained above	
Temperature (TL)	183°C
Time (t∟)	60-150 sec
Peak Temperature (T <sub>p</sub> )	240°C (+/-5°C)
Time within $5^{\circ}$ C of Actual Peak Temperature ( $t_p$ )	10-30 sec

Table 4: Reflow Profile Details, according to IPC/JEDEC J-STD-020D.1

#### 5.2. Multi-sensor integration

Mechanical coupling between drive frequencies of several sensors can affect performance at system level, for example within Inertial Measurement Units. Customer has to take care of such coupling during system design and validation.

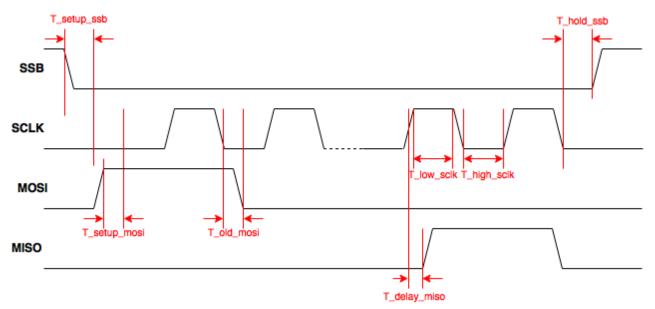
#### 5.3. Traceability

Label integrity has been validated with Vigon<sup>®</sup> and IPA. For other chemical treatment, the label integrity is not guaranteed.

## 6. Digital SPI interface

#### 6.1. Electrical and Timing Characteristics

The device acts as a slave supporting only SPI "mode 0" (clock polarity CPOL=0, clock phase CPHA=0).



#### Figure 20: SPI timing diagram

Symbol	Parameter	Condition	Unit	Min	Тур	Max
Electrical charact	teristics					
VIL	Low level input voltage		VDD	0		0.1
VIH	High level input voltage		VDD	0.8		1
VOL	Low level output voltage	ioL=0mA (Capacitive Load)	V		GND	
VOH	High level output voltage	ioH=0mA (Capacitive Load)	V		VDD	
Rpull_up	Pull-up resistor	Internal pull-up resistance to VDD	kΩ		100	
Rpull_down	Pull-down resistor	Internal pull-down resistance to GND	kΩ		-	
Timing paramete	ers					
Fspi	SPI clock input frequency	Maximal load 25pF on MOSI or MISO	MHz		0.2	8
T_low_sclk	SCLK low pulse		ns	62.5		
T_high_sclk	SCLK high pulse		ns	62.5		
T_setup_mosi	MOSI setup time		ns	10		
T_hold_mosi	MOSI hold time		ns	5		
T_delay_miso	MISO output delay	Load 25pF	ns			40
T_setup_ssb	SSB setup time		Tsclk	1		
T_hold_ssb	SSB hold time		Tsclk	1		

Table 5: SPI timing parameters

The MISO pin is kept in high impedance when the SSB level is high, which allows sharing the SPI bus with other components.

IMPORTANT NOTE: It is forbidden to keep SPI pads at a high level while VDD is at 0V due to ESD protection diodes and buffers.

#### 6.2. SPI frames description

The SPI frames used for the communication through the SPI Register are composed of an instruction followed by arguments. The SPI instruction is composed of 1 byte, and the arguments are composed of 2, 4 or 8 bytes, depending on the cases, as can be seen in Table 6 below.



Figure 21: SPI Message Structure

Instruction	Argument	Meaning
0x50	0x00000000 (n=4)	Read Angular Rate
0x54	0x0000 (n=2)	Read Temperature
0x58	0x00000000 (n=4)	Advanced commands.
0x78	0xXXXXXXXX (n=8)	See Section 6.5 for more details.
0x7C	0xXXXX (n=2)	uetalls.

Table 6: Authorized SPI commands

#### 6.3. Angular rate readings

From the 32-bits (4 bytes) frame obtained after the "Read Angular Rate" instruction, the 24-bits word of angular rate data (RATE) must be extracted as shown below in Figure 22.

DRY and ST are respectively the "data ready" and "self-test" bits.

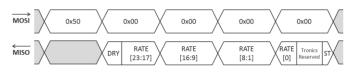


Figure 22: Angular rate reading frames and data organization

#### 6.3.1. Angular rate (RATE) output

The 24-bit gyro output is coded in two's complement (Table 7).

- If the temperature compensation is not enabled (GOUT\_SEL=1), then the user should perform scale factor measurements.
- If the temperature compensation of the angular rate output is enabled (default case), dividing the

24-bit value by a factor **10 000** results in the angular rate in  $^{\circ}$ /s, as shown in Table 7.

-350.0000	°/s	⇔	1100 1010 1001 1000 0010 0000			
 -300.0000 	°/s	⇔	1101 0010 0011 1001 0100 0000			
-0.0002	°/s	$\Leftrightarrow$	1111 1111 1111 1111 1111 1110			
-0.0001	°/s	$\Leftrightarrow$	1111 1111 1111 1111 1111 1111			
0.0000	°/s	$\Leftrightarrow$	0000 0000 0000 0000 0000 0000			
+0.0001	°/s	$\Leftrightarrow$	0000 0000 0000 0000 0000 0001			
+0.0002	°/s	$\Leftrightarrow$	0000 0000 0000 0000 0000 0010			
 +300.0000	°/s	⇔	0010 1101 1100 0110 1100 0000			
 +350.0000	°/s	$\Leftrightarrow$	0011 0101 0110 0111 1110 0000			
Table 7: Co	Table 7: Conversion table for calibrated angular rate output					

#### 6.3.2. Data Ready (DRY) bit

The Data Ready bit is a flag which is raised when a new angular rate data is available. The flag stays raised until the new data is read.

#### 6.3.3. Self-Test (ST) bit

The ST bit raises a flag (1 logic) at the same frequency as the angular rate output data rate indicating whether if the sensor is properly operating (i.e. whether the drive loop control provides stable drive oscillations amplitude).

The self-test procedure is running in parallel to the main functions of the sensor.

The ST data is also available on the pin 15. This pin is set to VDD when the sensor is working properly.

#### 6.4. Temperature readings

The temperature data is an unsigned integer, 12-bits word (TEMP). It must be extracted from the 2 bytes of read data, as shown below in Figure 23.

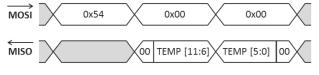


Figure 23: Temperature reading frames and data organization

By default the temperature sensor is *not* factory-calibrated (TOUTSEL=0).

#### 6.5. Advanced use of SPI registers

SPI registers can also be used to access the System register or the MTP (Multi-Time-Programmable memory).

#### 6.5.1. R/W access to the System Registers

**IMPORTANT NOTE:** Modifications to the system registers are **reversible**. Modified registers will *not* be restored after a RESET. There is no limitation to the number of times the system registers can be modified.

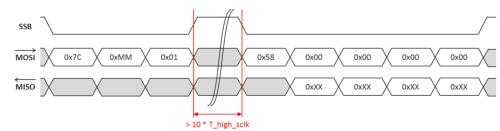


Figure 24: Sequence of instructions to READ address 0xMM of the system registers

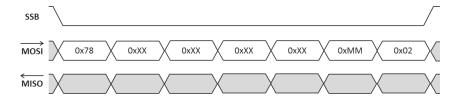


Figure 25: Sequence of instructions to WRITE '0xXXXXXXXX' to address '0xMM' of the system registers

#### 6.5.2. R/W access to the MTP

<u>IMPORTANT NOTE</u>: Modifications to the MTP are **non-reversible**. Modified parameters will be restored, even after a RESET, and previous values of the MTP cannot be accessed anymore. The maximum number of times the MTP can be written depends on the address:

- 7 times for the angular rate calibration coefficients (see Section 7 for more details)
- Only 1 time for all the other coefficients, including the temperature sensor calibration coefficients.

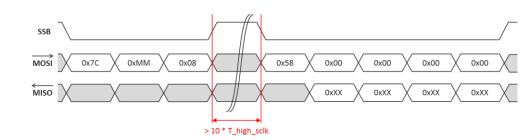


Figure 26 : Sequence of instructions to READ address 0xMM of the MTP

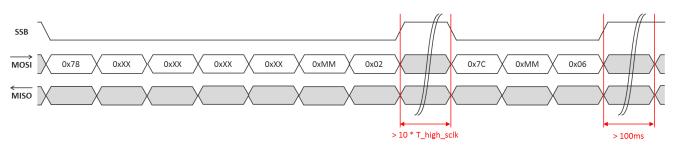


Figure 27: Sequence of instructions to WRITE data '0xXXXXXXXX' to address '0xMM' of the MTP

#### 6.5.3. Useful Sensor Parameters

The instructions given in Sections 6.5.1 and 6.5.2 can be used to read and/or to modify the sensor's useful parameters given in Table 8 below.

Parameter	Address M (System Register & MTP)	Bits	Encoding	Meaning
Sensor Identif	ication			
UID	0x00	[30:1]	Tronics reserved	Sensor 'Unique Identification' number
Temperature	output compens	ation		
TOUT_SEL	0x09	2*	0 **	Disable the calibrated temperature output
			1	Enable the calibrated temperature output
0	0x04	[27:16] *	0x000 **	Offset calibration of temperature sensor
			See section 8	
G	0x04	[13:2] *	0x800 **	Gain calibration of temperature sensor
			See section 8	
Angular rate o	output compensation	ation		
GOUT_SEL	0x02	27 *	0**	Enable the calibrated angular rate output
			1	Disable the calibrated angular rate output
SF2	0x2E	[31:16] *	See Table 9	Scale Factor 2 <sup>nd</sup> order coefficient (calibrated angular rate)
B2	0x2E	[15:0] *	See Table 9	Bias 2 <sup>nd</sup> order coefficient (calibrated angular rate)
B1	0x2F	[29:0] *	See Table 9	Bias 1 <sup>st</sup> order coefficient (calibrated angular rate)
B0	0x30	[29:0] *	See Table 9	Bias constant coefficient (calibrated angular rate)
SF1	0x31	[29:0] *	See Table 9	Scale Factor 1 <sup>st</sup> order coefficient (calibrated angular rate)
SFO	0x32	[29:0] *	See Table 9	Scale Factor constant coefficient (calibrated angular rate)
TMID	0x33	[19:0] *	See Table 9	Mid-temperature calibration point
MTPSLOTNB	0x02	[15:8] *	0b0000000	Unprogrammed part
			0b0000001 **	Programmed once, 7 slots remaining
			0b0000011	Programmed twice, 6 slots remaining
			0b01111111	Programmed 7 times, 1 slot remaining
			0b11111111	Programmed 8 times, no slot remaining

**Table 8: Useful parameters information** 

Notes:

\* The other bits at those addresses shall remain unchanged. Please make sure that you write them without modification!

\*\* Default Value

## 7. Angular rate calibration procedure

#### 7.1. Algorithm overview

After filtering, the raw angular rate sensor output is temperature compensated based on the on-chip temperature sensor output and the stored temperature compensation parameters.

#### 7.1.1. Angular rate output calibration model

The formula below models the link between raw and compensated angular rate outputs:

 $RATE[^{\circ}/s] = \frac{RATE_{COMP}[LSB]}{SF_{setting}[LSB/^{\circ}/s]} = \frac{RATE_{RAW}[LSB] - BIAS[LSB]}{SF[LSB/^{\circ}/s]}$ 

where:

- RATE is the angular rate output converted in °/s;
- RATECOMP is the calibrated angular rate output;
- SF<sub>setting</sub> is the constant conversion factor from LSB to °/s for the calibrated angular rate output. Default value for this parameter is SF<sub>setting</sub> = 10 000;
- RATE<sub>RAW</sub> is the raw data angular rate output;
- **BIAS** is a polynomial (2<sup>nd</sup> degree) temperaturevarying coefficient to model the sensor's bias temperature variations;
- **SF** is a polynomial (2<sup>nd</sup> degree) temperature-varying coefficient to model the sensor's Scale Factor temperature variations.

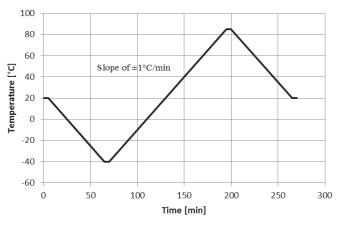


Figure 28: Recommended Temperature profile for calibration

#### 7.1.2. Recommended procedure

- 1. Set GOUT\_SEL to 1 in the System Registers (disable the calibration)
- Place the sensor on a rate table in a thermal chamber and implement temperature profile according to Figure 28<sup>1</sup>
- 3. Perform continuous acquisition of the angular rate output with the following pattern:
  - Rest position (0°/s input) to evaluate the BIAS parameter
  - + 300°/s input then -300°/s input to evaluate the SF parameter<sup>2</sup>
- 4. Calculate the coefficients of BIAS and SF polynomials:

$$BIAS = \sum_{i=0}^{2} b_i (T_{RAW} - T_{MID})^i$$
$$SF = \sum_{i=0}^{2} sf_i (T_{RAW} - T_{MID})^i$$

where

- T<sub>RAW</sub> is the raw output of the temperature sensor multiplied by 256;
- T<sub>MID</sub> is the mid-value of T<sub>RAW</sub>;
- b<sub>0</sub> to b<sub>2</sub> are the 3 coefficients of BIAS polynomial;
- sf<sub>0</sub> to sf<sub>2</sub> are the 3 coefficients of SF polynomial.
- 5. Convert  $T_{MID}$ ,  $b_i$  and  $sf_i$  parameters to their binary values according to Table 9 below:

Parameter	Value (decimal)	Format
SF2	$sf_2 \cdot 2^{55} / SF_{setting}$	signed 2's comp
SF1	$sf_1 \cdot 2^{46} / SF_{setting}$	signed 2's comp
SF0	$sf_0 \cdot 2^{27} / SF_{setting}$	signed 2's comp
B2	b <sub>2</sub> . 2 <sup>39</sup>	signed 2's comp
B1	b1 . 2 <sup>35</sup>	signed 2's comp
B0	bo	signed 2's comp
TMID	T <sub>MID</sub>	unsigned

Table 9: Angular rate calibration parameters

<sup>1</sup> Temperature profile can be adapted to be in line with customer applications

<sup>2</sup> Rate applied can be adapted to be in line with customer applications

# **tronics**

#### 7.2. Programming of the new coefficients

**IMPORTANT NOTE:** The following steps are **non-reversible**. The previous values of the coefficients will not be accessible anymore. The temperature compensation coefficients can be re-programmed up to 7 additional times on the IC.

The programming procedure consists in three major steps:

- Checking the available MTP slot status
- Programming the coefficients
- Updating the available MTP slot status

An overview of the procedure is given in Figure 29.

#### 7.2.1. Checking the MTP slot status

The first step is to check the number of remaining MTP slots (MTPSLOTNB), in other words, checking how many times the chip has been programmed before.

The detailed information of MTPSLOTNB register content is given in Table 8. The sequence of instructions to read the register is given in Figure 26.

The MTP slot number (MTPSLOTNB) re-programming iteration is given in the following table:

Iteration	Correspondence	MTP number	
		Value	Binary
0	Unprogrammed part	0	0000000
1	Programmed once	1*	0000001
2	Programmed twice	3	00000011
3		7	00000111
4		15	00001111
5		31	00011111
6		63	00111111
7		127	01111111
8	Cannot be further	255	11111111
	programmed		

Table 10: MTPSLOTNB iterations

\* Default value

#### 7.2.2. Programming the coefficients

This step describes the procedure for programming the calculated coefficients (temperature compensation of angular rate output). The programming procedure is:

- 1. Write SF2 in the system register
- 2. Write B2 in the system register
- 3. Program SF2 & B2 in the MTP
- 4. Write SF1 in the system register
- 5. Program SF1 in the MTP
- 6. Write SF0 in the system register
- 7. Program SF0 in the MTP
- 8. Write B1 in the system register

- 9. Program B1 in the MTP
- 10. Write B0 in the system register
- 11. Program B0 in the MTP
- 12. Write TMID in the system register
- 13. Program TMID

The detailed SPI commands are given in section 6.5. The detailed information about each coefficient is given in Table 8.

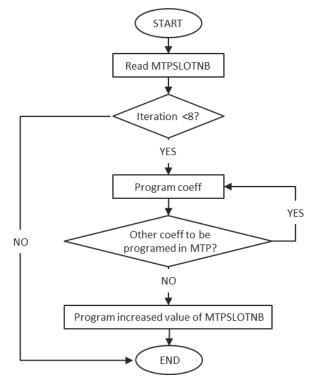


Figure 29 Procedure to program new calibration parameters

#### 7.2.3. Updating MTP slot status

This section describes the procedure for programming the updated status of the MTP slots.

# If this step is not performed properly, the new compensation coefficients will not be effective.

- 1. Read the MTPSLOTNB as described in section 6.5.2.
- 2. Increment MTPSLOTNB according to Table 10.
- 3. Write the updated MTPSLOTNB in the system register.
- 4. Program the updated MTPSLOTNB in the MTP.
- 5. After a reset, the new coefficients will be available.

#### 7.3. Switch to uncompensated data output

To optimize the thermal compensation of the angular rate output, it is possible to disable the on-chip compensation and use the uncompensated (raw) output to perform an external thermal compensation.

To switch the output to uncompensated data, the procedure is described on section 6.5, by modifying the GOUT register described on Table 8.

## 8. Temperature Sensor Calibration Procedure

The temperature output of GYPRO2300 sensors is *not* factory-calibrated, since only the relative temperature output is needed to perform temperature compensation of the angular rate output. However, it is possible to perform a first-order polynomial calibration of the temperature sensor, in order to output the absolute temperature information.

This section shows how to get and store temperature calibration parameters for the temperature output.

#### 8.1. Temperature sensor calibration model

The formula below models the link between raw and calibrated temperature output:

$$T[^{\circ}C] = \frac{T_{COMP}[LSB]}{GAIN_{setting}[LSB/^{\circ}C]} = \frac{GAIN \cdot T_{RAW}[LSB] + OFFSET[LSB]}{GAIN_{setting}[LSB/^{\circ}C]}$$

where:

- T is the output temperature converted in °C;
- T<sub>COMP</sub> is the calibrated temperature output;
- GAIN<sub>setting</sub> is the constant conversion factor from LSB to °C for the calibrated temperature output. This gain is set to 20LSB/°C to provide an output resolution of 0,1°C;
- T<sub>RAW</sub> is the raw data temperature output;
- OFFSET is a constant coefficient to tune the offset;
- **GAIN** is a constant coefficient to tune gain.

The **OFFSET** and **GAIN** parameters will be computed and written in the ASIC as per the following calibration procedure.

#### 8.2. Recommended Procedure

- 1. Check that TOUT\_SEL = 0. If not, set it to 0 in the System Registers.
- 2. Measure the temperature output with at least 2 temperature points  $T_1$  and  $T_2$ .

3. Calculate the GAIN and OFFSET coefficients according to formula above.

$$\text{GAIN} = \text{GAIN}_{setting} \cdot \frac{\text{T1}_{ABS}[^{\circ}\text{C}] - \text{T2}_{ABS}[^{\circ}\text{C}]}{\text{T1}_{RAW}[\text{LSB}] - \text{T2}_{RAW}[\text{LSB}]}$$

 $OFFSET = GAIN_{setting} \cdot T1_{ABS}[^{\circ}C] - GAIN \cdot T1_{RAW}[LSB]$ 

where:

- T1<sub>ABS</sub> is the absolute temperature of T<sub>1</sub> in °C;
- T2<sub>ABS</sub> is the absolute temperature of T<sub>2</sub> in °C;
- T1<sub>RAW</sub> is the raw output temperature of T<sub>1</sub> in LSB;
- T2<sub>RAW</sub> is the raw output temperature of T<sub>2</sub> in LSB;
- 4. Convert GAIN and OFFSET to their binary values according to Table 11 below:

Parame	ter Value (decimal)	Format	
G	GAIN . 2 <sup>09</sup>	Unsigned	
0	OFFSET	Unsigned	
Table 11. Tawa anatuma calibuatian wanawatawa			

Table 11: Temperature calibration parameters

- 5. [<u>Optional step:</u> Write GAIN and OFFSET into the System Registers and repeat step 2. to check the accuracy of the new calibration.]
- 6. Write GAIN and OFFSET into the MTP according to instructions of Section 6.5.2. Meanwhile, set TOUT\_SEL to 1 during this step, so that the new calibration parameters are effective after a RESET.



# 9. Device Identification / Ordering information

#### 9.1. Device identification

GYPRO2300 tracking information is accessible on the label, as shown in the next figure.

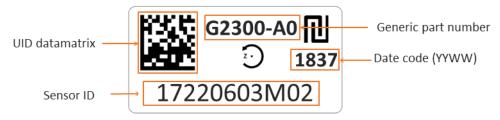
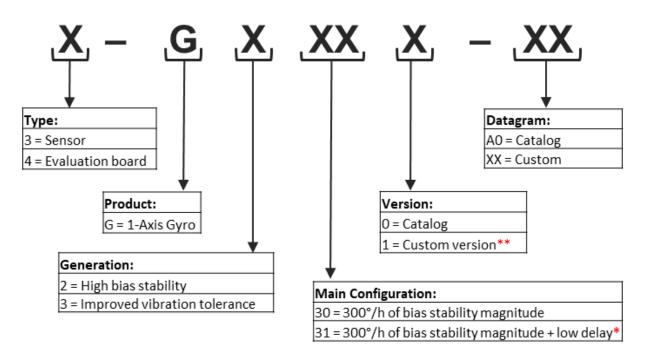


Figure 30: GYPRO2300 label.

9.2. Ordering information



#### Figure 31 Ordering information

\* For second 2<sup>nd</sup> generation only

\*\* Custom version or specific requirement can be address upon request.

Product	Ordering code
GYPRO2300	3-G2300-A0
GYPRO2300LD	3-G2310-A0
GYPRO3300	3-G3300-A0
GYPRO2300-EVB2	4-G2300-A0
GYPRO2300LD-EVB2	4-G2310-A0
GYPRO3300-EVB2	4-G3300-A0

## **10.Internal construction and Theory of Operation**



Figure 32 : Inner view of the package, showing the MEMS and IC

GYPRO series is using the dominant architecture for high performance MEMS gyro, namely the "Tunning fork or dual mass" design.

In details, each sensor consists in a MEMS transducer and an integrated circuit (IC) packaged in a 30-pins Ceramic Leadless Chip Carrier Package.

The sensing element (MEMS die), which is located on the left part of the Figure 32, is manufactured using Tronics' waferlevel packaging technology based on micro-machined thick single crystal silicon. The MEMS consists of two coupled substructures subjected to linear anti-phase vibrations. The structures are vacuumed at the wafer-level providing high Qfactor in the drive mode. The drive system is decoupled from the sense system in order to reduce feedback from sense motion to drive electrodes. The drive anti phase vibration is sustained by electrostatic comb drives. The sense anti phase vibration resulting from Coriolis forces is counter balanced by electrostatic forces. Differential detection and actuation are used for both drive and sense systems and for each substructure, keeping two identical structures for efficient common mode rejection.

The integrated circuit (IC), which is located on the right part of the Figure 32, is designed to interface the MEMS sensing element. It includes ultra-low noise capacitive to voltage converters (C2V) followed by high resolution voltage digitization (ADC) for both drive and sense paths. Excitation voltage required for capacitance sensing circuits is generated on the common electrode node. 1-bit force feedbacks (DAC) are used for both drive and sense system actuation.

The choice for the implemented close-loop architecture based on a Sigma-Delta principle is particularly well adapted as it brings the following key advantages:

1) Sigma-Delta is well suited for low-frequency signals. Noise shaping principle rejects quantization noise in high frequency bands. 2) Simplicity of hardware implementation. Oversampling concept allows significant design relaxation of the analog detection chain signal resolution. Additionally the voltage reference used for actuation force feedback is also of simple implementation as it is a 1-bit D/A converter, thus simplifying its design.

3) Linearization of the electrostatic forces thanks to the Sigma-delta principle (through force averaging) furthermore reduces non-linearity overall and more importantly its evenorder terms, which result in rectification error.

4) Sigma-Delta signal output is inherently a digital signal, thus suppressing the need for costly high resolution A/D converter.

The digital part implements digital drive and sense loops, demodulates, decimates and processes the gyro output based on the on-chip temperature sensor output. The system controller manages the interface between the SPI registers, the system register and the non-volatile memory (OTP). The nonvolatile memory provides the gyro settings, in particular the coefficients for angular rate sensor temperature compensation. On power up, the gyro settings are transferred from the OTP to the system registers and output data are available in the SPI registers. The angular rate sensor output and the temperature sensor output are available in the SPI registers. The SPI registers are available through the SPI interface (SSB, SCLK, MOSI, MISO). The self-test is available on the external pins ST.

The "References" block generates the required biasing currents and voltages for all blocks as well as the low-noise reference voltage for critical blocks.

The "Power Management" block manages the power supply of the sensor from a single 5V supply between the VDD and GND pins. It includes a power on reset as well as an external reset pin (RSTB) to start or restart operation using default configuration. An enable pin (EN) with power-down capability is also available.

The sensor is powered with a single 5V DC power supply through pins VDD and GND. Although the sensor contains three separate VDD pins, the sensor is supplied by a single 5V voltage source. It is recommended to supply the three VDD pins in a star connection with appropriate decoupling capacitors. Regarding the sensor grounds, all the GND pins are internally shorted. The GND pins redundancy is used for multiple bonds in order to reduce the total ground inductance. It is therefore recommended to connect all the GND pins to the ground.

## **11.Available Tools and Resources**

The following tools and resources are available on the GYPRO® product page of our website or upon request.

Item	Description			
Documentation & technical notes				
The second s	GYPRO <sup>®</sup> product line - Flyer			
A A A A A A A A A A A A A A A A A A A	<b>GYPRO® product</b> – Technical note External filtering for Gypro2300LD and Gypro3300			
The second se	<b>GYPRO® product</b> – Technical note GYPRO MTBF Methodology			
Mechanical tools				
	GYPRO2300 – 3D model			
Evaluation kit				
	GYPRO2300-EVB2 – Evaluation board Evaluation board for GYPRO2300, compatible with Arduino Yun_rev2			
	Evaluation Board – User manual			
100 100 100	<b>Evaluation Kit</b> – Quick start guide			
	Evaluation Tool – Software user manual			
The same is a second seco	<b>GYPRO<sup>®</sup> Evaluation Tool</b> – Tutorial Installation and programming of the Evaluation kit			
	<b>GYPRO<sup>®</sup> Evaluation Tool</b> – Tutorial Software			
	Evaluation Tool – Software			
	Evaluation Tool – Arduino Firmware			